

Improvement of the Output Mode Purity of a Complex-Cavity Resonator for a Frequency-Tunable Sub-THz Gyrotron

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Abstract—One way to improve the mode selection in terahertz gyrotrons is the use of stepped cavity resonators that satisfy the resonance condition for two modes with the same azimuthal but different radial indices. Such sectioned resonators, or complex cavities, were analyzed in detail by several authors, recently and in the past. One common problem for such designs is the excitation of unwanted modes by both the stepped cavity structure and the cavity input and output tapers. Mode conversion becomes especially a severe problem for higher order resonant modes in highly oversized resonators. In this paper we introduce ways to minimize the unwanted mode conversion in both, the radial step of the cavity as well as in the tapered sections of the cavity and the uptaper to collector and output window. Frequency tuning between 391.5 and 393.5 GHz has been achieved by operation at different axial mode numbers of the mode pair $TE_{8,4}/TE_{8,5}$. Due to their intrinsic higher output mode purity, the so-called anti-phase modes of the complex cavity have been chosen for this output mode purity study.

Index Terms—Gyrotron, Frequency Tuning, Complex Cavity Resonator, Output Mode Purity

I. INTRODUCTION

COMPLEX stepped cavities can improve the mode selectivity in very high frequency low-power gyrotrons [1-3]. Such type of mode converting cavities have been investigated in much detail in the past [4-12]. It is well known, that in the case of sub-THz gyrotrons the mechanical precision of the relative ratio of the two cavity radii must be in the order of 10^{-4} , which is approx. $1/Q$, where Q is in the order of the quality factors of the two cavity sections. However, even if this mechanical accuracy is achieved, there still remains the question of spurious mode excitation by the cavity step and the output taper of the second section. The present paper deals with this problem and proposes methods to improve the output mode purity of a frequency-tunable sub-THz gyrotron with complex cavity. Frequency tuning over 2 GHz is achieved by operation on different axial cavity modes ($q = 1, 2, 3, 4, 5$). In all cases an output mode purity ≥ 92.5 % of the gyrotron could be demonstrated. When the frequencies of the partial modes in the two cavity sections are close, their superposition causes

formation of normal modes (in-phase and anti-phase mode) [3]. For the in-phase mode, the phase is nearly constant at the junction of the two sections, while for the anti-phase mode it exhibits an abrupt sharp change of π . The features of these two normal modes were studied very carefully in [3], where it was shown that the diffractive quality factors Q_{dif} for the anti-phase modes are much larger than those for the in-phase ones. With the increase of the axial index q , Q_{dif} of the anti-phase modes decreases approximately as q^{-2} , whereas, for the in-phase modes Q_{dif} decreases more slowly. The Ohmic quality factor Q_{ohm} of both normal modes are nearly the same and they slightly increase with q . The $TE_{8,4}$ mode slightly penetrates into the second cavity section where it propagates as a far from cutoff traveling forward wave. For the in-phase mode, this penetration is much stronger than for the anti-phase one, and the amplitude of the $TE_{8,4}$ mode is nearly the same or even larger as the amplitude of outgoing $TE_{8,5}$ wave. This is the reason for the lower Q_{dif} of the in-phase mode. Since we here in the present paper deal with output mode purity, we selected the anti-phase mode for the further considerations.

II. MULTI-MODE CAVITY ANALYSIS

AS an example we took the cavity dimensions given in [3] for a complex gyrotron cavity with resonant coupling of the $TE_{8,4,q}/TE_{8,5,q}$ modes. The wall radii of the resonant section of the cavity are $R_1 = 2.588$ mm and $R_2 = 2.9974$ mm, respectively. The length of both resonant sections L_1 and L_2 is 17.5 mm and the down-tapering and up-tapering angles are $\theta_1 = 2^\circ$ and $\theta_2 = 3^\circ$, respectively. The cavity geometry is plotted in Fig. 1. The cavity was simulated using a resonator code based on the scattering matrix formalism [13] taking into account all propagating and additional evanescent modes. In particular, for the following calculations, we took into account up to 12 TE- and 12 TM-Modes. The results of these simulations for several axial modes are plotted in Fig. 2. The resulting resonance frequencies and diffractive quality factors are given in Table I and are in excellent agreement with the values obtained in [3]. However, the output mode purity ($TE_{8,5}$ mode content at the cavity output) is rather poor and might therefore prohibit the use of such a cavity in an experimental gyrotron.

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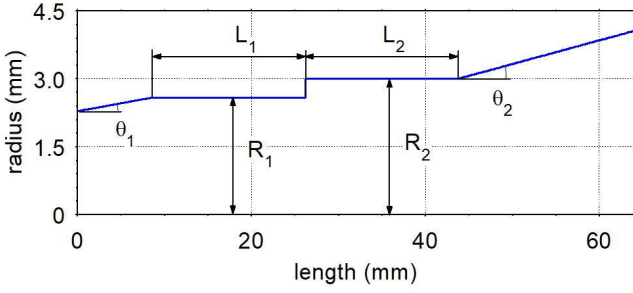


Fig. 1. Radial contour of the complex $TE_{8,4,q}/TE_{8,5,q}$ mode gyrotron cavity from [3].

TABLE I

COMPARISON OF THE CALCULATED RESONANCE FREQUENCIES AND DIFFRACTIVE QUALITY FACTORS (f_{res}^* AND Q_{dif}^*) OF THE FIRST FIVE ANTI-PHASE AXIAL MODES FROM [3] WITH THE CORRESPONDING VALUES (f_{res} AND Q_{dif}) CALCULATED HERE USING THE MULTI-MODE SCATTERING MATRIX FORMALISM. THE OBTAINED OUTPUT MODE PURITIES ARE GIVEN IN THE LAST COLUMN

q	f_{res}^* (GHz)	Q_{dif}^*	f_{res} (GHz)	Q_{dif}	$TE_{8,5}$ (%)
1	391.462	45002	391.474	44247	75.14
2	391.716	10363	391.725	10106	79.98
3	392.139	4650	392.149	4488	75.81
4	392.727	2467	392.738	2430	77.27
5	393.483	1640	393.495	1662	82.49

III. IMPROVED STEP-TYPE COUPLING

CIRCULAR waveguide modes with the same azimuthal but different radial indices can be effectively coupled by applying a simple abrupt step in the waveguide diameter. To investigate the influence of the radial step in the cavity given in Fig. 1 on the output mode purity, the output radius of a simple stepped waveguide with a fixed input radius $R_1 = 2.588$ mm and 100 % $TE_{8,4}$ as input mode was varied at a constant frequency of $f = 391.471$ GHz. The length of the straight waveguide sections at both ends of the abrupt step is 10 mm. As wall material we chose high purity copper at room temperature with a conductivity of $5.6 \cdot 10^7$ S/m. The resulting reflected and transmitted modal power as a function of the output radius R_2 is plotted in Fig. 3. Whereas the reflection occurs in the pure $TE_{8,4}$ mode, step-type coupling to modes with higher radial indices occurs in forward direction. As can be seen from Fig. 3 (e), there is also conversion to many modes with lower radial indices, which is most severe at radii R_2 where the higher order modes are close to cutoff. Unfortunately, these are exactly the radii where gyrotron oscillations take place. Fig. 4 shows the calculated reflection and transmission as a function of frequency for the waveguide step of the complex gyrotron cavity plotted in Fig. 1 [3] with $R_1 = 2.588$ mm and $R_2 = 2.9974$ mm. As can be clearly seen, strong mode conversion in this complex cavity containing a simple abrupt radial step is practically unavoidable. Introduction of a linear 45° up-taper between the two sections [2,7] does not change the situation essentially.

Irises in cavities of sub-THz gyrotrons have been applied mainly to enhance the diffractive quality factor of the resonators [2,13-15]. Here we use a radial stub to improve the output mode

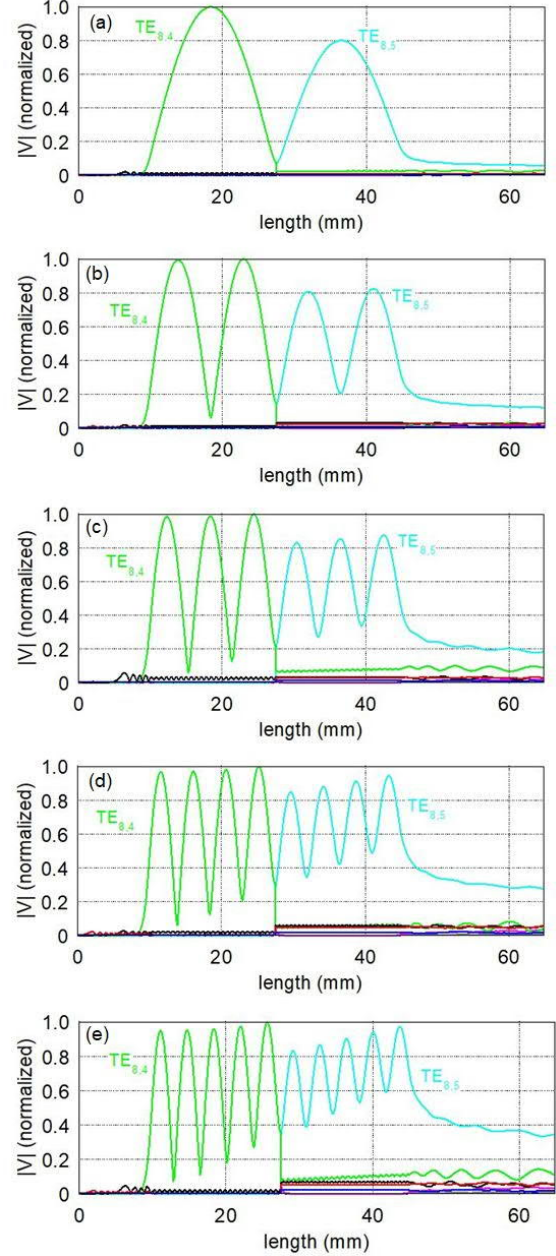


Fig. 2. Calculated axial profiles of the first five anti-phase axial modes in the complex $TE_{8,4,q}-TE_{8,5,q}$ mode gyrotron cavity from [3]. (a) $q = 1$, (b) $q = 2$, (c) $q = 3$, (d) $q = 4$, (e) $q = 5$.

purity. Into the complex cavity of Fig. 1 [3] we introduce at the end of the first cavity section with radius R_1 a larger radial step to an intermediate radius R_i followed by a short intermediate section with length L_i and then a small radial step down to the radius R_2 of the second cavity section (Fig. 5). The dimensions of the stub were optimized numerically. Figure 6 gives the calculated output mode purity of the $TE_{8,5}$ mode in the larger waveguide section with $R_2 = 2.9974$ mm for $TE_{8,4}$ as input mode in the smaller cavity section with $R_1 = 2.588$ mm. There is a wide range of parameters R_i and L_i providing the lowest undesired mode conversion with about 81 % coupling to the $TE_{8,5}$ mode. We chose the smallest possible dimensions with $R_i = 3.15$ mm and $L_i = 0.42$ mm, indicated by the black triangle in Fig. 6. The electric field distributions along the waveguide

radial step from $R_1 = 2.588$ mm to $R_2 = 2.9974$ mm without stub and with optimized stub at the frequency of 391.471 GHz are plotted in Fig. 7. Obviously the function of the stub is to reconvert most of the spectrum of the excited propagating lower order and evanescent higher order modes $TE_{8,n}$ and $TM_{8,n}$ into the desired $TE_{8,5}$ mode. Figure 8 gives the calculated reflection and transmission of the radial step with optimized stub as a function of frequency. The $TE_{8,5,q}$ -mode contents in the transmitted mode spectrum of the radial step with and without radial stub are compared in Table II for the resonance frequencies of the first five axial modes, showing the substantial improvement of mode purity by the optimized stub.

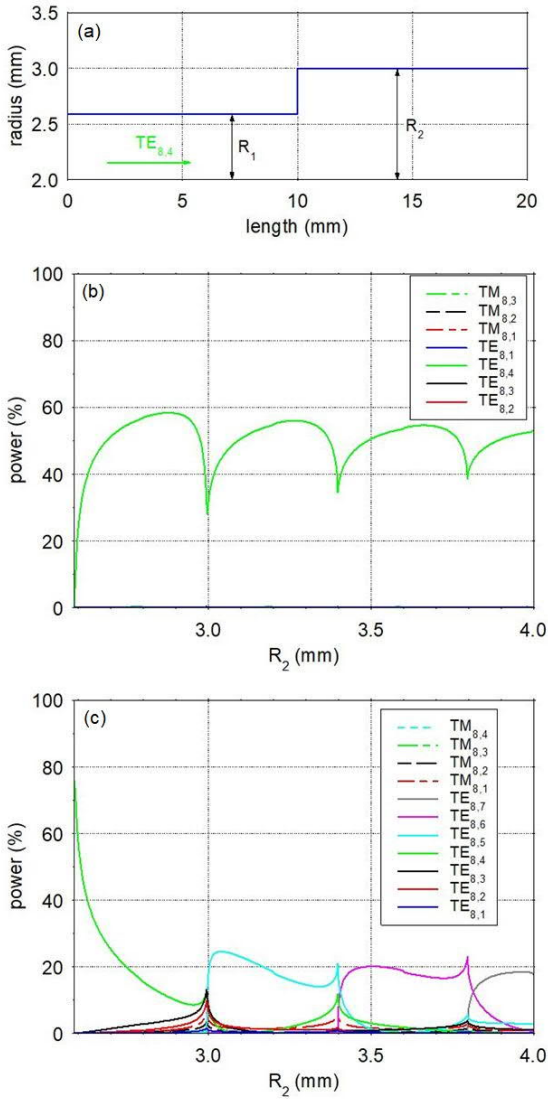


Fig. 3. Abrupt waveguide step with constant input radius $R_1 = 2.588$ mm fed by the $TE_{8,4}$ mode at a frequency of 391.471 GHz, radial contour (a), calculated reflection (b) and transmission/mode conversion (c) as a function of the output radius R_2 . In (b) the corresponding dependencies for other modes than $TE_{8,4}$ are not visible due to the selected scale.

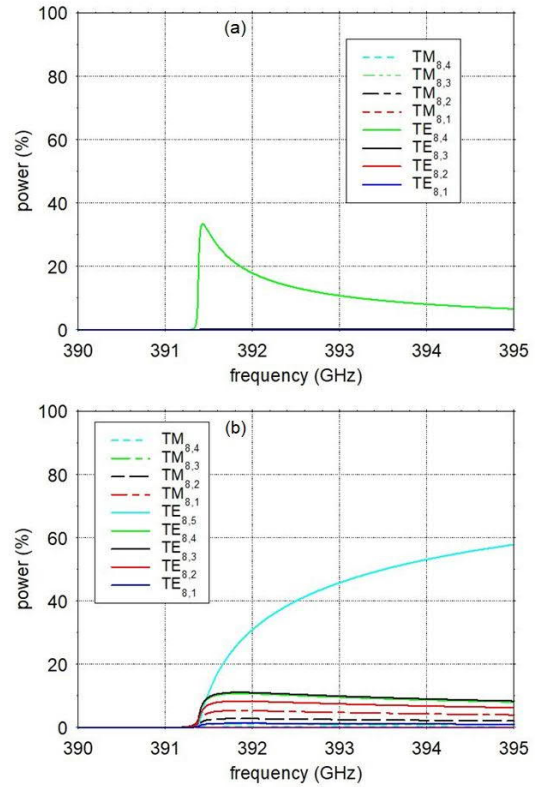


Fig. 4. Calculated reflection (a) and transmission/conversion (b) of the $TE_{8,4}$ mode at a waveguide step with input radius $R_1 = 2.588$ mm and output radius $R_2 = 2.9974$ mm as a function of frequency. In (a) the corresponding dependencies for other modes than $TE_{8,4}$ are not visible due to the selected scale.

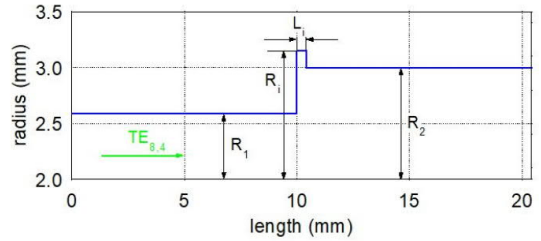


Fig. 5. Radial contour of waveguide step with radial stub and input radius $R_1 = 2.588$ mm and output radius $R_2 = 2.9974$ mm.

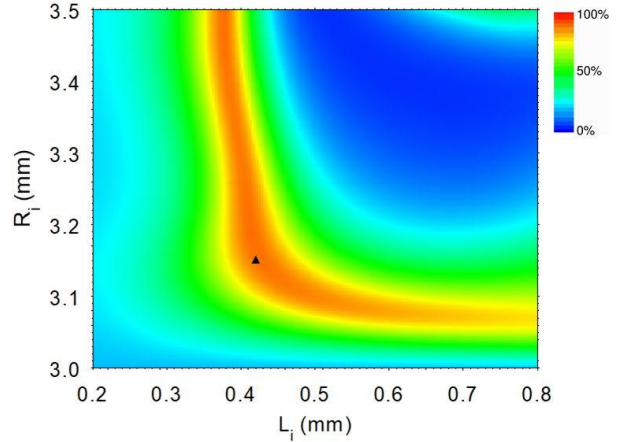


Fig. 6. Calculated output mode purity (percentage of the $TE_{8,5}$ mode power in the output mode spectrum) of a waveguide step as a function of the radial stub dimensions (L_1 , R_1) for an input radius of $R_1 = 2.588$ mm and an output radius of $R_2 = 2.9974$ mm at $f = 391.471$ GHz. The smallest possible stub dimensions with optimal coupling are indicated by the black triangle.

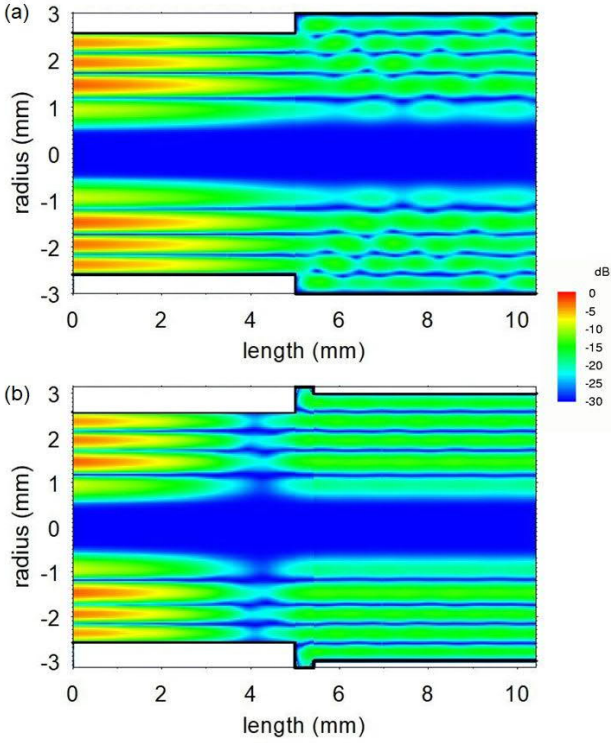


Fig. 7. Calculated azimuthal electric field distribution at a waveguide step from $R_1 = 2.588$ mm to $R_2 = 2.9974$ mm without (a) and with optimized stub with $R_1 = 3.15$ mm and $L_i = 0.42$ mm with 100 % $TE_{8,4}$ as input mode at the frequency of $f = 391.471$ GHz.

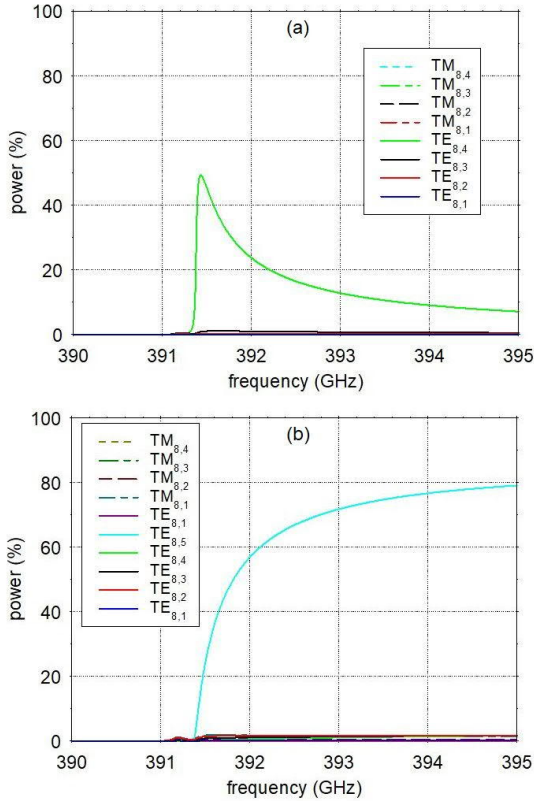


Fig. 8. Calculated reflection (a) and transmission/conversion (b) of the radial step from $R_1 = 2.588$ mm to $R_2 = 2.9974$ mm with optimized stub with $R_1 = 3.15$ mm and $L_i = 0.42$ mm as a function of frequency.

TABLE II
COMPARISON OF THE MODE CONVERSION OF THE FIRST FIVE ANTI-PHASE AXIAL $TE_{8,5,q}$ MODES AT A SIMPLE RADIAL STEP FROM $R_1 = 2.588$ mm TO $R_2 = 2.9974$ mm WITH A STUB-LOADED STEP WITH $R_1 = 3.15$ mm AND $L_i = 0.42$ mm

q	f_{res} (GHz)	$TE_{8,5}$ (%), simple step	$TE_{8,5}$ (%), step with stub
1	391.471	21.19	80.95
2	391.721	35.65	88.65
3	392.138	45.79	90.94
4	392.720	53.16	91.70
5	393.465	58.72	91.84

IV. COMPLEX CAVITY WITH RADIAL STUB

WE replaced the simple abrupt step in the complex cavity of Ref. [3] with the radial stub described in Section III. All other dimensions of the cavity remain the same as in Fig. 1. The geometry of the modified complex cavity with stub is plotted in Fig. 9. Figure 10 shows the calculated axial field profiles for the first five anti-phase axial modes. The comparison with results from the original cavity design in Table III and Fig. 2 shows a significant improvement with respect to the achieved output mode purity. The diffractive quality factor of the complex cavity also increased due to the radial stub, which is due to both, the slightly increased reflection at the stub compared to the simple abrupt step and the improved mode purity in the cavity.

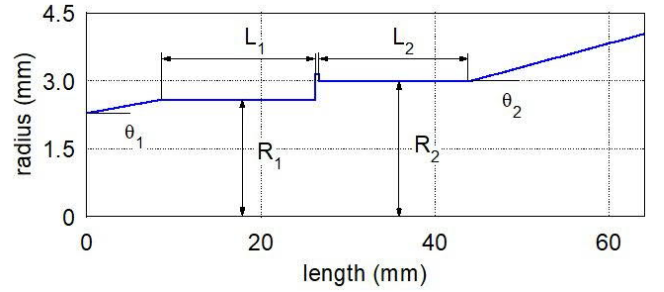


Fig. 9. Radial contour of the complex $TE_{8,4,q}/TE_{8,5,q}$ mode gyrotron cavity with intermediate radial stub.

TABLE III
CALCULATED RESONANCE FREQUENCIES f_{res} , DIFFRACTIVE QUALITY FACTORS Q_{dif} AND OUTPUT MODE PURITIES OF THE FIRST FIVE ANTI-PHASE AXIAL $TE_{8,5,q}$ MODES FOR THE CAVITY WITH RADIAL STUB

q	f_{res} (GHz)	Q_{dif}	$TE_{8,5}$ (%)
1	391.471	57232	87.51
2	391.721	14868	92.55
3	392.138	6761	89.98
4	392.720	3839	90.77
5	393.465	2597	93.92

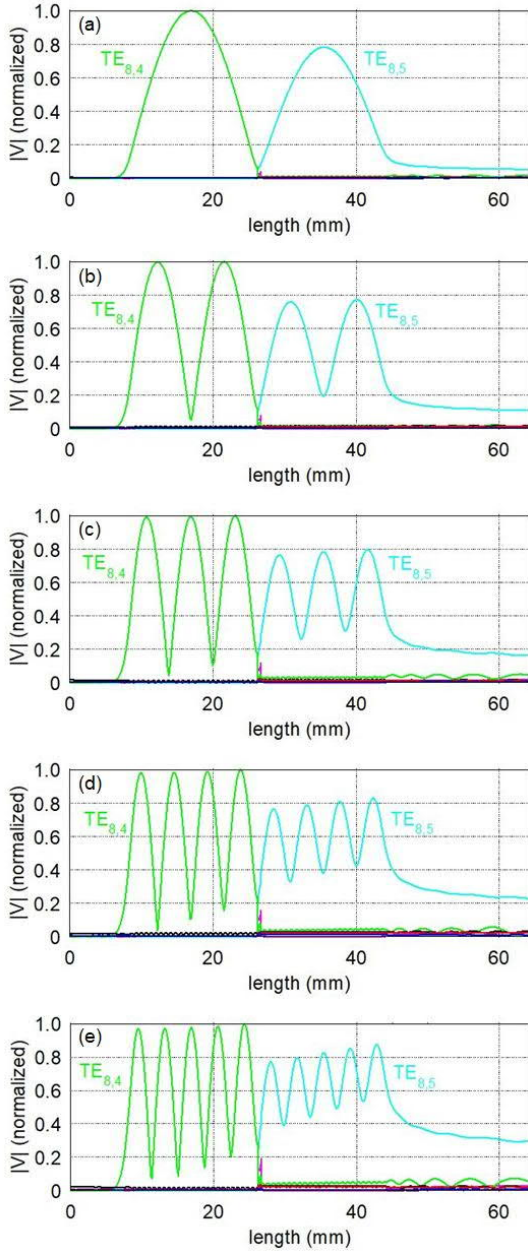


Fig. 10. Calculated axial profiles of the first five anti-phase axial modes in the complex $TE_{8.4,q}$ - $TE_{8.5,q}$ mode cavity with intermediate stub. (a) $q = 1$, (b) $q = 2$, (c) $q = 3$, (d) $q = 4$, (e) $q = 5$.

V. OPTIMIZED NONLINEAR COMPLEX CAVITY WITH STUB

THE final output radius of experimental gyrotrons is usually much larger than that given in Ref. [3]. In order to determine the output mode purity of a gyrotron the whole output taper section has to be taken into account. In case of a simple linear output taper design, which very often is used in experimental devices, this can lead to large fractions of parasitic modes in the output radiation of the gyrotron. To demonstrate this, we assumed a final output window radius of 9.0 mm, similar to gyrotron FU VA described in Ref. [14]. The output mode purity can be increased significantly with a rounded nonlinear cavity design [13]. A schematic rounding is shown in Fig. 11.

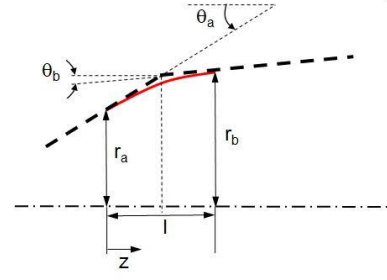


Fig. 11. Rounding between two adjacent circular waveguide tapers.

The main parameters are the angles of the adjacent tapers θ_a and θ_b and the length of the rounded section l . The radial contour of a rounded section with length l , matching the gradients of the adjacent tapers is then given by (1).

$$r(z) = r_a + z \cdot \left[\tan \theta_a + z \cdot \left(\frac{3 \cdot A - B}{2} + z \cdot \frac{B - A}{2 \cdot l} \right) \right] \quad (1)$$

with

$$A = \frac{r_b - r_a - \tan \theta_a \cdot l}{l^2}$$

$$B = \frac{\tan \theta_b - \tan \theta_a}{2 \cdot l}$$

Based on the original linear complex cavity (Fig. 1), roundings were introduced both at the upstream (l_1) and downstream tapers (l_2 , l_3 , see also the schematic drawing in Fig. 12).

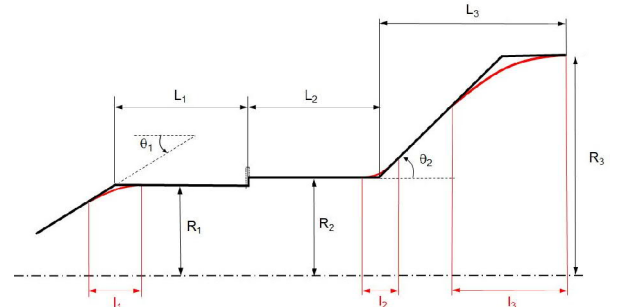


Fig. 12. Geometrical parameters of the nonlinear cavity design with roundings.

A numerical optimization using the scattering matrix code revealed a maximum output mode purity and a similar diffractive quality factor compared to the linear complex cavity for the following parameters of the nonlinear cavity with radial stub and an output radius of 9.0 mm: $\theta_1 = 2^\circ$, $\theta_2 = 4^\circ$, $l_1 = 2$ mm, $l_2 = 4$ mm, $l_3 = 95$ mm, $L_3 = 139.5$ mm. The contour of the optimized nonlinear cavity with radial stub is plotted in Fig. 13.

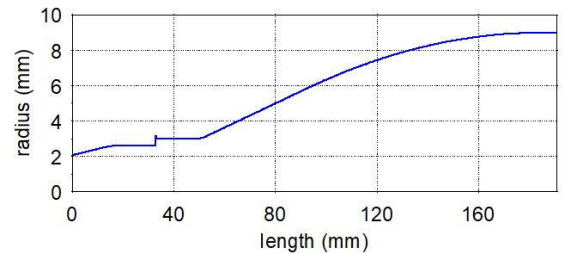


Fig. 13. Radial contour of the optimized nonlinear complex cavity with intermediate stub.

Table IV compares the performance of linear input and output tapers with the nonlinear tapers of the optimized cavity design (Fig. 13). It contains the calculated mode purity of the $TE_{8,4,q}$ modes reflected at the input taper and of the $TE_{8,5,q}$ mode transmitted through the output taper for $q = 1-5$. For both tapers the nonlinear design leads to higher mode purities compared to the linear contour. There is substantial improvement especially for the output taper.

TABLE IV
COMPARISON OF THE CALCULATED REFLECTION OF THE $TE_{8,4}$ MODE AT THE DOWN-TAPER AND THE TRANSMISSION OF THE $TE_{8,5}$ MODE THROUGH THE UP-TAPER BETWEEN THE LINEAR (FIG. 1) AND THE NONLINEAR (FIG. 13) CAVITY DESIGN AT THE RESONANCE FREQUENCIES OF THE COMPLEX CAVITY FOR THE FIRST FIVE ANTI-PHASE AXIAL $TE_{8,4,q}/TE_{8,5,q}$ MODES

q	f_{res} (GHz)	$TE_{8,4}$ (%) down-taper		$TE_{8,5}$ (%) up-taper	
		linear	rounded	linear	rounded
1	391.475	99.15	99.99	79.37	99.65
2	391.739	98.68	99.99	79.70	99.62
3	392.175	98.33	99.99	80.06	99.58
4	392.783	98.18	99.99	80.53	99.52
5	393.557	98.28	99.99	80.91	99.46

Finally, we compare the output mode purity of the original linear complex cavity of Ref. [3] in combination with a linear output up-taper from $R_2 = 2.9974$ mm to $R_3 = 9.0$ mm with the optimized nonlinear design (Fig. 13). Figure 14 shows the axial field profiles of the first 5 axial modes in the complete optimized nonlinear geometry. The output mode purities of the linear and the nonlinear gyrotron designs at the resonance frequencies are compared in Table V.

The very poor output mode purity of the linear gyrotron design results from both, mode conversion at the step in the cavity and in the linear output taper. With the addition of the intermediate radial stub in the complex cavity in combination with an optimized nonlinear cavity profile the overall mode conversion could be successfully reduced to just a few percent for all the considered 5 axial cavity modes.

TABLE V
COMPARISON OF THE CALCULATED RESONANCE FREQUENCIES, AND OUTPUT MODE PURITIES OF THE COMPLEX CAVITY OF REF. [3] WITH LINEAR TAPER FROM RADIUS $R_2 = 2.9974$ mm TO $R_3 = 9.0$ mm AND CONSTANT TAPER ANGLE OF $\theta_2 = 3^\circ$ WITH THE COMPLEX CAVITY DESIGN WITH STUB AND NONLINEAR TAPERS FOR THE FIRST FIVE ANTI-PHASE AXIAL $TE_{8,4,q}/TE_{8,5,q}$ MODES

q	Cavity with simple step and linear input and output tapers			Cavity with stub and nonlinear input and output tapers		
	f_{res} (GHz)	Q_{dif}	$TE_{8,5}$ (%)	f_{res} (GHz)	Q_{dif}	$TE_{8,5}$ (%)
1	391.472	44097	67.71	391.475	47974	96.36
2	391.725	10141	67.41	391.739	12349	95.54
3	392.149	4533	61.68	392.175	5614	95.01
4	392.738	2493	74.01	392.783	3037	92.48
5	393.495	1733	55.09	393.557	1921	96.66

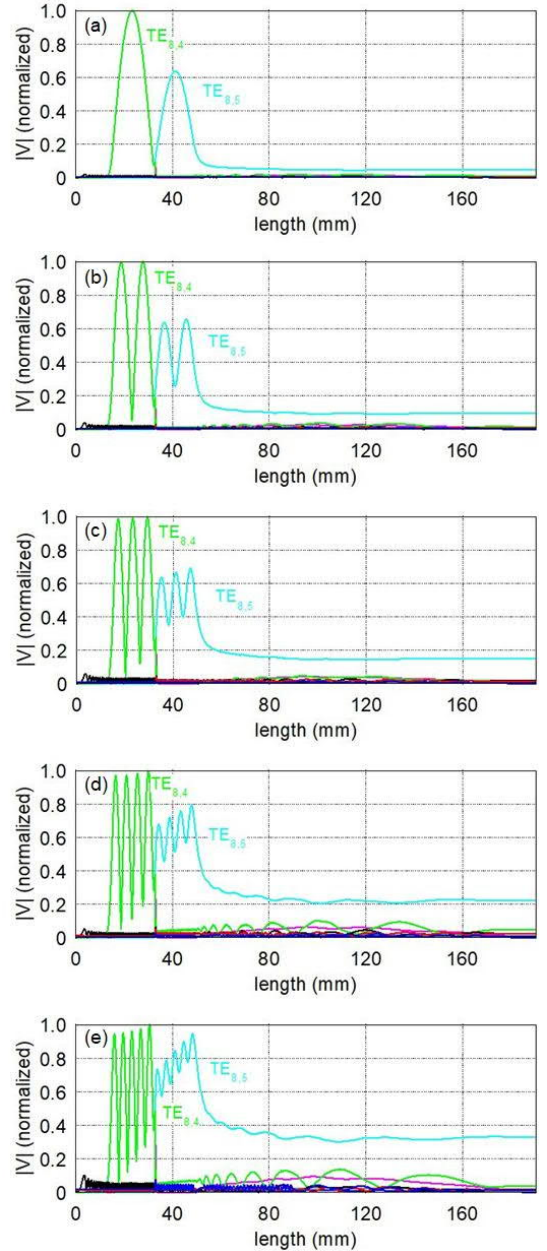


Fig. 14. Calculated axial profiles of the first five anti-phase axial modes of the optimized complex cavity with intermediate stub and nonlinear q output taper from $R_2 = 2.9974$ mm to $R_3 = 9.0$ mm. (a) $q = 1$, (b) $q = 2$, (c) $q = 3$, (d) $q = 4$, (e) $q = 5$.

VI. CONCLUSIONS

A complex $TE_{8,4,q}/TE_{8,5,q}$ -mode cavity for a frequency-tunable sub-THz gyrotron has been analyzed using a multimode resonator code based on the scattering matrix formalism. It has been shown that a simple radial step between the two cavity sections causes significant conversion to parasitic modes, even when the ratio of the radii of the two sections is precisely correct. The introduction of an additional groove (stub) between the two cavity sections removes this mode conversion almost completely for all the five axial modes at the different frequencies from 391.5 to 393.5 GHz. Roundings were introduced both at the upstream and

downstream tapers of the stepped cavity. By additionally replacing the linear uptaper to collector and output window with an optimized nonlinear taper, in all cases a gyrotron output mode purity of $\geq 92.5\%$ could be demonstrated. Since the rounding at the cavity output and the output taper angle have been chosen in such a way that the quality factors of the original and the improved stepped cavity are approximately the same, we expect similar electronic efficiencies of the two versions. However, this has to be proven with electron-beam-wave interaction calculations in a following paper.

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